

ECE 2025, Georgia Tech
solution to PS # 9

9.1

a) $x_a[n] = 5s[n-1] + 3^n u[n-2]$

$$x_a[n] = 5s[n-1] + 3^2 (3)^{n-2} u[n-2]$$

$$X_a(z) = 5z^{-1} + 9 \frac{z^{-2}}{1-3z^{-1}} = \frac{5z^{-1}-6z^{-2}}{1-3z^{-1}}$$

b) $x_b[n] = 2\left(-\frac{1}{2}\right)^n u[n] + 3\left(-\frac{1}{3}\right)^n u[n]$

$$X_b(z) = 2 \frac{1}{1+\frac{1}{2}z^{-1}} + 3 \frac{1}{1+\frac{1}{3}z^{-1}} = \frac{5 + \frac{13}{6}z^{-1}}{(1+\frac{1}{2}z^{-1})(1+\frac{1}{3}z^{-1})}$$

c) $x_c[n] = n u[n]$

$$u[n]*u[n] = \sum_{k=0}^{\infty} u[k] u[n-k] = \sum_{k=0}^n 1 = n+1 \quad \text{for } n \geq 0$$

thus $u[n]*u[n] = (n+1) u[n] \rightarrow x_c[n] = u[n]*u[n] - u[n]$

$$\rightarrow X_c(z) = \frac{1}{(1-z^{-1})^2} - \frac{1}{1-z^{-1}} = \frac{z^{-1}}{(1-z^{-1})^2}$$

9.2

a) $y(z) = -0.9 \bar{z}^1 y(z) + x(z) - \bar{z}^1 x(z)$

$$\rightarrow y(z) (1 + 0.9 \bar{z}^1) = x(z) (1 - \bar{z}^1)$$

$$\rightarrow H(z) = \frac{Y(z)}{X(z)} = \frac{1 - \bar{z}^1}{1 + 0.9 \bar{z}^1} = \frac{1}{1 + 0.9 \bar{z}^1} - \frac{\bar{z}^1}{1 + 0.9 \bar{z}^1}$$

$$\rightarrow h[n] = (0.9)^n u[n] - (-0.9)^{n-1} u[n-1]$$

b) $x[n] = u[n-2] \rightarrow X(z) = \frac{\bar{z}^2}{1 - \bar{z}^1}$

$$Y(z) = H(z) X(z) \rightarrow Y(z) = \frac{1 - \bar{z}^1}{1 + 0.9 \bar{z}^1} \cdot \frac{\bar{z}^2}{1 - \bar{z}^1} = \frac{\bar{z}^2}{1 + 0.9 \bar{z}^1}$$

$$\rightarrow y[n] = (-0.9)^{n-2} u[n-2]$$

c) $bb = [1 \ -1] ;$

$$aa = [1 \ 0.9] ;$$

$$xx = [0 \ 0 \ ones(1, 9)] ;$$

$$yy = filter(bb, aa, xx) ; \quad index = 0 : 1 : (length(yy)-1);$$

stem(yy, index)

(2)

9.3

$$a) H(e^{j\hat{\omega}}) = \frac{1 - (e^{j\hat{\omega}})^{-1}}{1 + 0.9(e^{j\hat{\omega}})^{-1}} = \frac{1 - e^{-j\hat{\omega}}}{1 + 0.9e^{-j\hat{\omega}}}$$

b) You may use matlab to plot this. However we can also approximately plot it using the pole-zero locations.

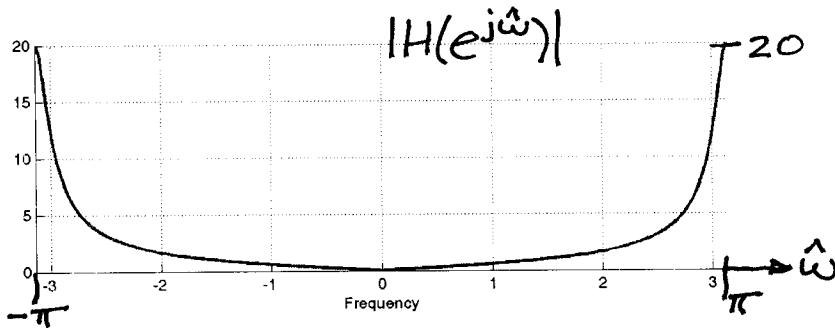
Note that $|H(e^{j\hat{\omega}})|^2 = \frac{1 - e^{-j\hat{\omega}}}{1 + 0.9 e^{-j\hat{\omega}}} \cdot \frac{1 - e^{j\hat{\omega}}}{1 + 0.9 e^{j\hat{\omega}}} = \frac{2(1 - \cos \hat{\omega})}{1.81 + 1.8 \cos \hat{\omega}}$

$H(z)$ has one zero at $z=1$ and a pole at $z=-0.9$

Thus, $|H(e^{j\hat{\omega}})| = 0$ at $\hat{\omega} = 0$ and $|H(e^{j\hat{\omega}})|$ is large at $\hat{\omega} = \pi$
(i.e., has a peak)

That is $|H(e^{j\pi})|^2 = \frac{2(2)}{1.81 - 1.8} = 400$

The plot is even symmetric w.r.t. $\hat{\omega}$.



$$c) H(e^{j0.8\pi}) = \frac{1 - e^{-j0.8\pi}}{1 + 0.9 e^{-j0.8\pi}} \stackrel{j0.45\pi}{\approx} 3.2 e^{j0.45\pi}$$

$$\Rightarrow y[n] = 320 \cos(0.8\pi n - 0.2\pi + 0.45\pi) \\ = 320 \cos(0.8\pi n + 0.25\pi)$$

9.4

Freq. Response A : should have a zero at $\hat{\omega} = 0$, i.e. $z = 1$.

Thus, S_3, S_4, S_6, S_8 are possible candidates.

However, $|H(e^{j\hat{\omega}})|$ at $\hat{\omega} = \pi$ for S_3, S_6, S_8 are $\frac{8}{3}, 4, \frac{8}{3}$, respectively. Hence, the correct answer is S_4 for which $|H(e^{j\hat{\omega}})| = 1.2$ at $\hat{\omega} = \pi$

Freq. Res. B : Must have zero at $z = -1$ (i.e., $\hat{\omega} = \pi$).

Possible answers are S_1, S_2, S_7 . However, only S_2 satisfies the condition $|H(e^{j0})| = H(z)|_{z=1} = 8$

Freq. Res. C : should have a complex conjugate zero (at $0 < \hat{\omega} < \frac{\pi}{2}$). Thus, possible candidate is S_5 .

All the other systems have either zero at $\omega = 0$ or $\hat{\omega} = \pi$.

Freq. Res. D : Must have zero at $z = -1$ ($\hat{\omega} = \pi$).

Possible answer: S_1, S_2, S_7 .

However, only S_1 satisfies the magnitude condition

$$\text{at } \hat{\omega} = 0 \quad |H(e^{j0})| = \frac{20}{6}$$

Freq. Res. E : Must have a zero at $z = 1$ and a zero at $z = j$ ($\hat{\omega} = \frac{\pi}{2}$). Only S_6 satisfies this.

Freq. Res. F : zero at $z = -1$ and $H(1) = 0.5$ only S_7 satisfies these conditions. (we also expect a zero close to unit circle at $0 < \hat{\omega} < 1$). ④

9.5

a) $H_3(z) :$ $Y(z) = a_1 z^{-1} Y(z) + a_2 z^{-2} Y(z) + b_0 V_2(z)$

$$H_3(z) = \frac{Y(z)}{V_2(z)} = \frac{b_0}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$a_1 = a_2 = 3 \quad b_0 = 4 \quad \rightarrow H_3(z) = \frac{4}{1 - 3z^{-1} - 3z^{-2}} = \frac{4z^2}{z^2 - 3z - 3}$$

roots of $z^2 - 3z - 3 = 0$

$$\rightarrow z = \frac{3}{2} \pm \frac{1}{2}\sqrt{21} \quad \text{since } H_3(z) \text{ has poles outside of}$$

unit circle, then the causal system is unstable.

b) $H(z) = H_1(z) H_2(z) H_3(z) = \frac{4z^2(z^2 + 5z^{-3})(1 + 3z^{-1})}{z^2 - 3z - 3}$

c) $H(z) = \frac{b_0 z^2 (z^2 + 5z^{-3})(1 + 3z^{-1})}{z^2 - a_1 z - a_2} = 10 z^{-(n_d - 1)} + 55 \frac{z^{-n_d}}{1 - \frac{1}{2}z^{-1}}$

$$\rightarrow \frac{b_0 (1 + 5z^{-1})(1 + 3z^{-1})}{z^2 - a_1 z - a_2} = \frac{z^{-(n_d - 1)} (10z - 5 + 55)}{z - \frac{1}{2}}$$

① left side should have also pole at $z = \frac{1}{2} \Rightarrow (\frac{1}{2})^2 - a_1(\frac{1}{2}) - \frac{a_2}{2} = 0$ ①

② one at the zeros of leftside should cancel with a pole of leftside, so that left and right become equivalent.
Note that right side has a zero at $z = -5$

This $(1 + 3z^{-1})$ should be cancelled $\Rightarrow (-3)^2 - a_1(-3) - a_2 = 0$ ②
solving ① and ② $\Rightarrow a_1 = -2.5 \quad a_2 = 1.5$

9.5 cont.

$$\frac{b_0 z^2 (z+5)(z+3)}{(z+3)(z - \frac{1}{2})} = \frac{10 z^{(n_d-1)} (z+5)}{z - \frac{1}{2}}$$

$$\Rightarrow b_0 = 10 \quad n_d = 3$$

9.6

s_1, s_2, s_3, s_4 are IIR $\Leftrightarrow \#4, \#2$

s_5, s_6, s_7, s_8 are FIR $\Leftrightarrow \#1, \#3$

Using locations of pole and zero for s_1, s_2, s_3, s_4

We can conclude : $s_3 \Leftrightarrow \#2$

$s_2 \Leftrightarrow \#4$

using the fact that s_5, s_7 have two poles and ^{two} zeros
while s_6, s_8 have 3 poles and 3 zeros, we

conclude : $s_5, s_7 \Leftrightarrow \#1$

$s_6, s_8 \Leftrightarrow \#3$

But using locations of zeros for plot 1 and
plot 3, we conclude :

$s_7 \Leftrightarrow \#1$

$s_6 \Leftrightarrow \#3$

9.7

FIR : $\delta, L, N \Leftrightarrow S_5, S_6, S_7, S_8$

IIR : $K, M, O \Leftrightarrow \underbrace{S_1, S_2, S_3, S_4}$

δ has four nonzero coeff. $\Leftrightarrow S_6, S_8$

Since S_6 takes only ± 1 values \Rightarrow

$$\delta \Leftrightarrow S_8$$

Similarly, N has four nonzero $\overset{\pm 1 \text{ value}}{\text{coeff}}$, thus

$$N \Leftrightarrow S_6$$

L has three nonzero values $\Leftrightarrow S_5, S_7$

Since S_5 takes ± 1 values, we conclude

$$L \Leftrightarrow S_7$$

using the inverse z-transform, we conclude that

$$K \Leftrightarrow S_4$$

$$M \Leftrightarrow S_3$$

$$O \Leftrightarrow S_2$$