

(12.1) P-11.7

(a) FROM THE TABLE:  $\frac{\sin(\omega_b t)}{\pi t} \longleftrightarrow u(\omega + \omega_b) - u(\omega - \omega_b)$

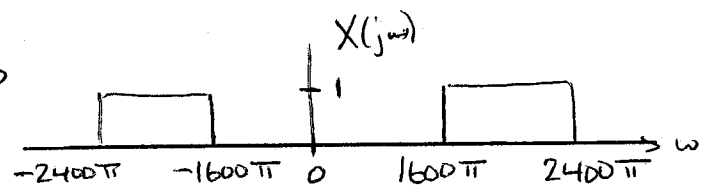
$$\Rightarrow \frac{10 \sin(200\pi t)}{\pi t} \longleftrightarrow 10 [u(\omega + 200\pi) - u(\omega - 200\pi)]$$

DIFFERENTIATION PROPERTY:  $\frac{d}{dt} y(t) \longleftrightarrow j\omega Y(j\omega)$

$$\Rightarrow \frac{d}{dt} \left[ \frac{10 \sin(200\pi t)}{\pi t} \right] \longleftrightarrow 10 j\omega [u(\omega + 200\pi) - u(\omega - 200\pi)]$$

(b)  $\frac{2 \sin(400\pi t)}{\pi t} \longleftrightarrow 2 [u(\omega + 400\pi) - u(\omega - 400\pi)]$

MODULATION PROPERTY:  $y(t) \cos(\omega_0 t) \longleftrightarrow \frac{1}{2} Y(j(\omega - \omega_0)) + \frac{1}{2} Y(j(\omega + \omega_0))$

$$\Rightarrow \frac{2 \sin(400\pi t)}{\pi t} \cos(2000\pi t) \longleftrightarrow$$


(c) FROM THE TABLE:  $\sum_{n=-\infty}^{\infty} \delta(t - nT) \longleftrightarrow \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{T} k)$

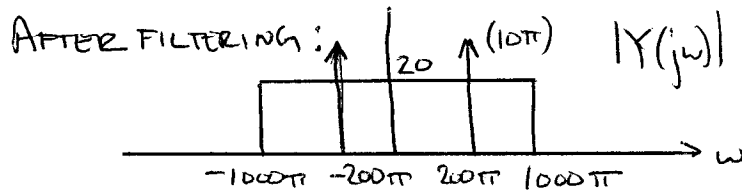
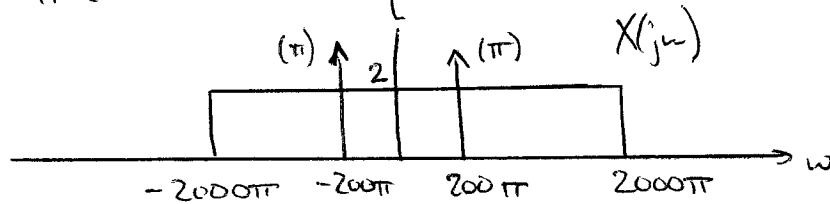
HERE,  $T = 10$ , SO

$$\sum_{n=-\infty}^{\infty} \delta(t - 10n) \longleftrightarrow \frac{\pi}{5} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{\pi k}{5})$$

12.2 P-11.13

$$(a) \cos(200\pi t) \leftrightarrow \pi \delta(\omega - 200\pi) + \pi \delta(\omega + 200\pi)$$

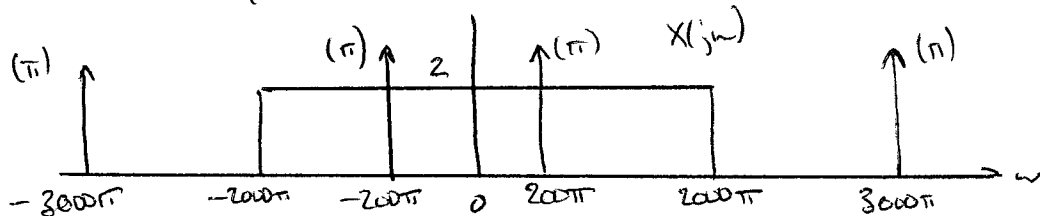
$$\frac{2 \sin(2000\pi t)}{\pi t} \leftrightarrow 2 \left[ u(\omega + 2000\pi) - u(\omega - 2000\pi) \right]$$



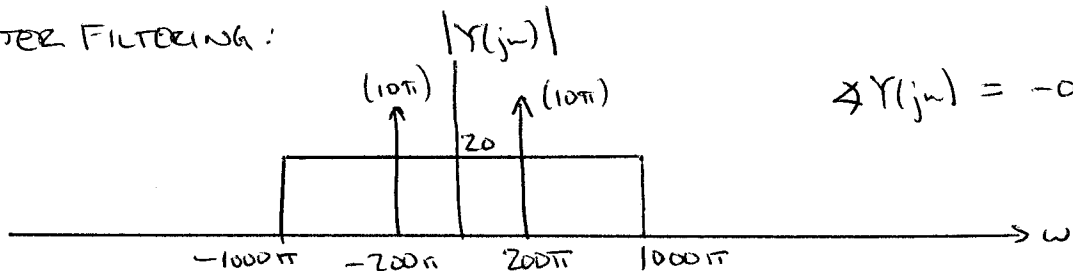
$$\star Y(j\omega) = -0.0025\omega$$

$$\Rightarrow y(t) = 10 \cos(200\pi(t - 0.0025)) + \frac{20 \sin[1000\pi(t - 0.0025)]}{\pi(t - 0.0025)}$$

$$(b) X(j\omega) = \pi \delta(\omega - 200\pi) + \pi \delta(\omega + 200\pi) + 2 \left[ u(\omega + 2000\pi) - u(\omega - 2000\pi) \right] + \pi \delta(\omega - 3000\pi) + \pi \delta(\omega + 3000\pi)$$



AFTER FILTERING:

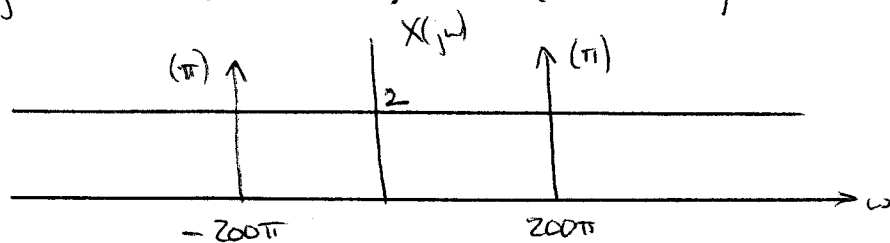


$$\star Y(j\omega) = -0.0025\omega$$

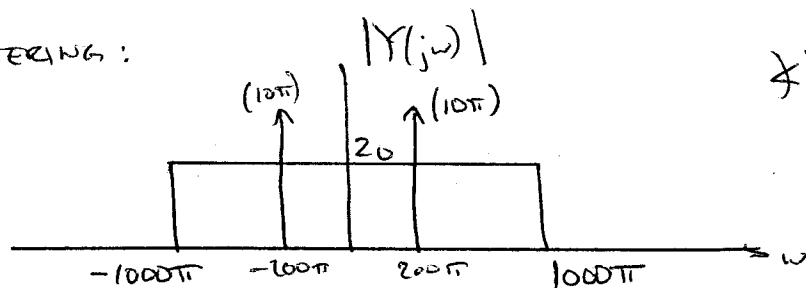
$$\Rightarrow y(t) = 10 \cos(200\pi(t - 0.0025)) + \frac{20 \sin[1000\pi(t - 0.0025)]}{\pi(t - 0.0025)}$$

(same as for (a))

$$(c) X(j\omega) = \pi \delta(\omega - 200\pi) + \pi \delta(\omega + 200\pi) + 2$$



AFTER FILTERING:



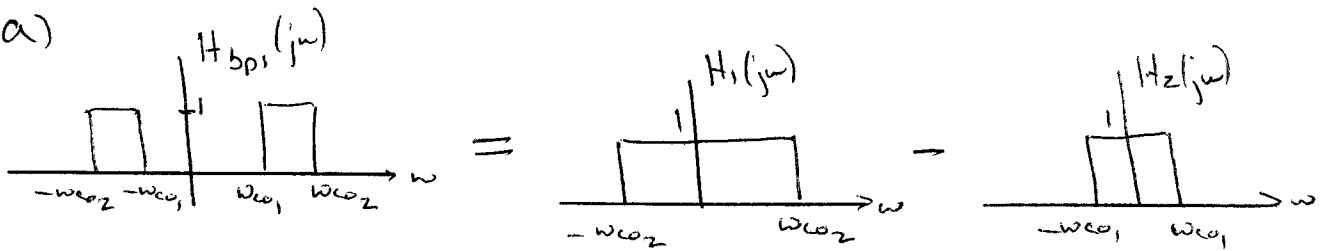
$$Y(j\omega) = -0.0025\omega$$

This is the same  $y(t)$  as for (a) and (b).

(c)  $H(j\omega)$  is an ideal lowpass filter with cutoff frequency  $1000\pi$  and a delay of  $0.0025$  s.

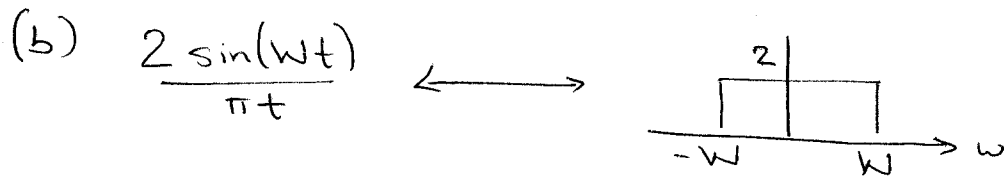
Because the input signals in each of these three cases is identical for  $|\omega| < 1000\pi$ , the output is identical.

12.3 (a)



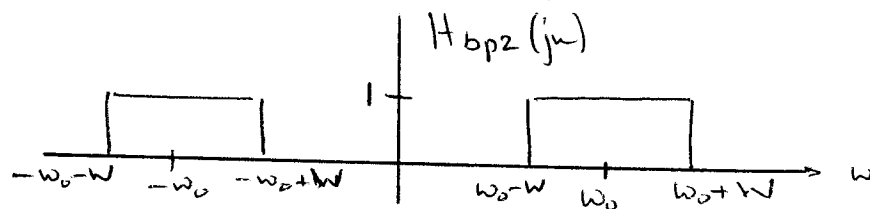
$$\frac{\sin(\omega_{c2} t)}{\pi t} \longleftrightarrow H_1(j\omega) \quad \text{and} \quad \frac{\sin(\omega_{c1} t)}{\pi t} \longleftrightarrow H_2(j\omega)$$

$$\Rightarrow h_{bp1}(t) = \frac{\sin(\omega_{c2} t)}{\pi t} - \frac{\sin(\omega_{c1} t)}{\pi t}$$



$$\text{and } x(t) \cos(\omega_0 t) \longleftrightarrow \frac{1}{2} X(j(\omega - \omega_0)) + \frac{1}{2} X(j(\omega + \omega_0))$$

Therefore,  $H_{bp2}(j\omega)$  has the graph:



(c)  $\omega_0$  is the average of  $\omega_{c1}$  and  $\omega_{c2}$ :

$$\omega_0 = \frac{\omega_{c1} + \omega_{c2}}{2}$$

$W$  is half the width of the filter passband:

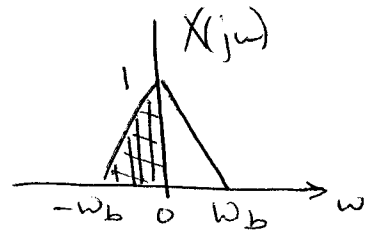
$$W = \frac{\omega_{c2} - \omega_{c1}}{2}$$

12.4

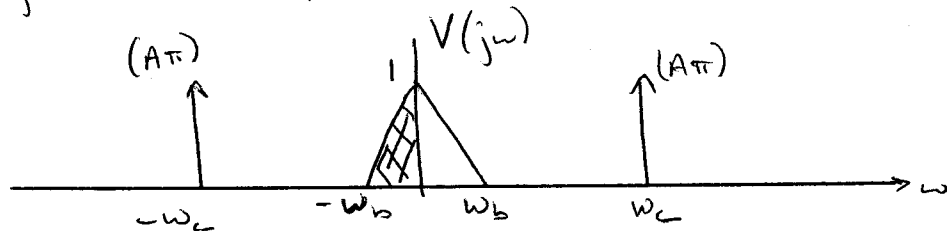
(a)  $v(t) = x(t) + A \cos(\omega_c t)$

$\Rightarrow V(j\omega) = X(j\omega) + A\pi \delta(\omega - \omega_c) + A\pi \delta(\omega + \omega_c)$

If  $X(j\omega)$  has the spectrum:



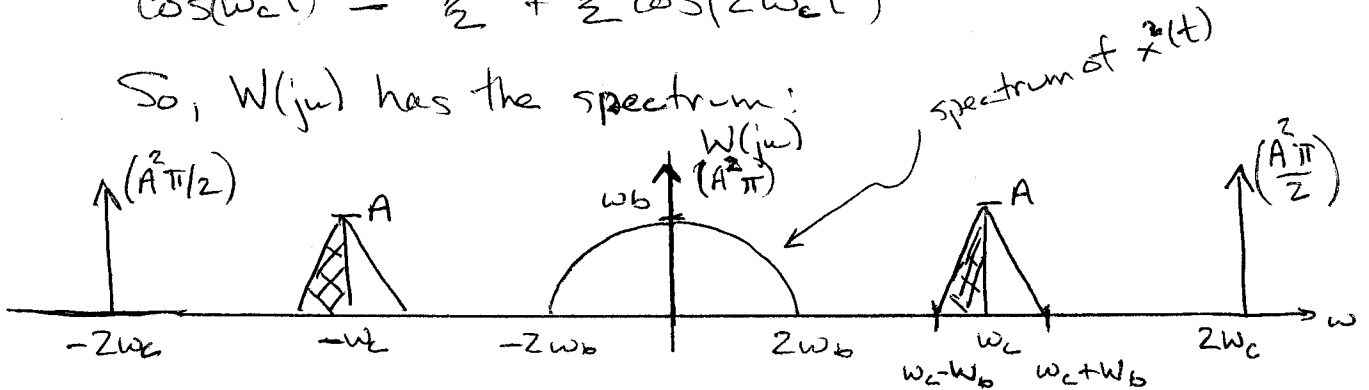
Then  $V(j\omega)$  has the spectrum:



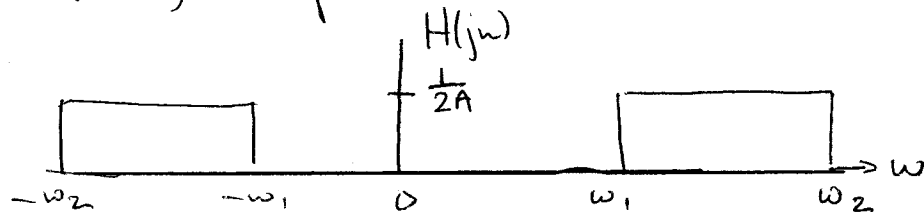
(b)  $w(t) = v^2(t) = x^2(t) + 2Ax(t)\cos(\omega_c t) + A^2\cos^2(\omega_c t)$

$\cos^2(\omega_c t) = \frac{1}{2} + \frac{1}{2}\cos(2\omega_c t)$

So,  $W(j\omega)$  has the spectrum:



(c) Need to filter  $w(t)$  so as to keep only the  $x(t)\cos(\omega_c t)$  component;



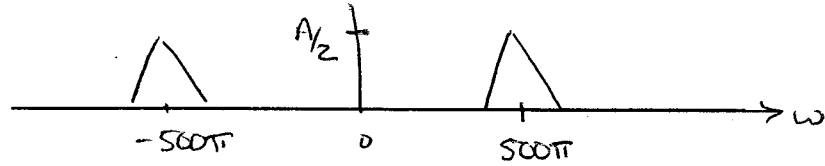
Need  $2\omega_b < \omega_1 < \omega_c - \omega_b$  and  $\omega_c + \omega_b < \omega_2 < 2\omega_c$ .

To avoid overlap in the spectral components of  $W(j\omega)$ , we need  $\omega_c > 3\omega_b$ .

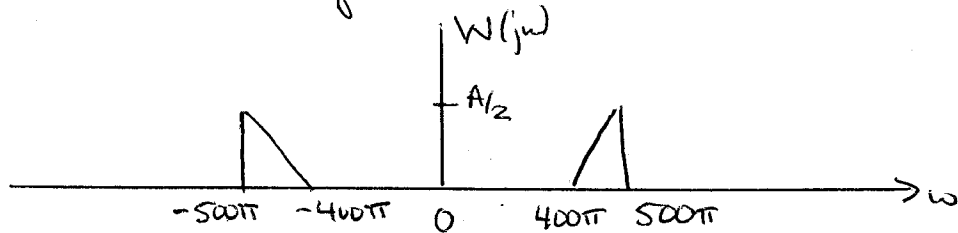
12.5

(a)

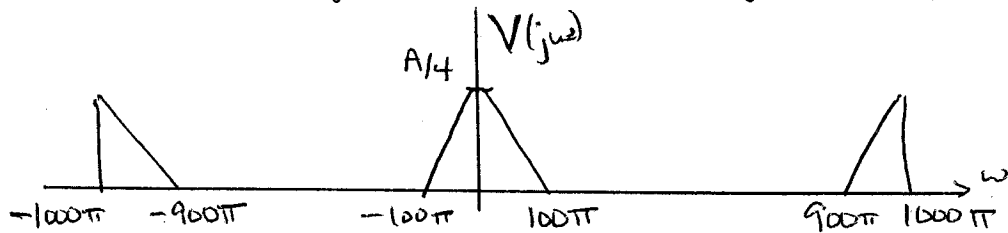
$$x(t) \cos(\omega_c t) \longleftrightarrow \frac{1}{2} X(j(\omega - \omega_c)) + \frac{1}{2} X(j(\omega + \omega_c))$$



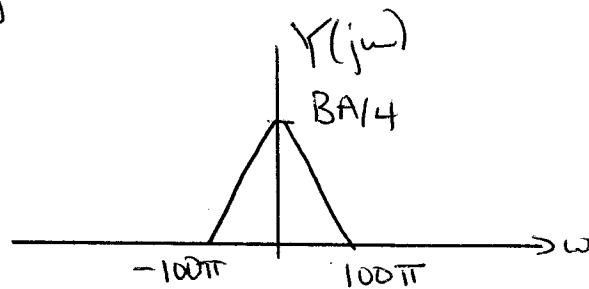
After lowpass filtering:



$$(b) V(j\omega) = \frac{1}{2} W(j(\omega - \omega_c)) + \frac{1}{2} W(j(\omega + \omega_c))$$



(c) For  $Y(j\omega)$  we have



If  $B=4$ , then  $Y(j\omega) = X(j\omega)$ .