

# ECE-2025 HW #7 Fall-2001

From Problem Set #6 Solution Sp. 99

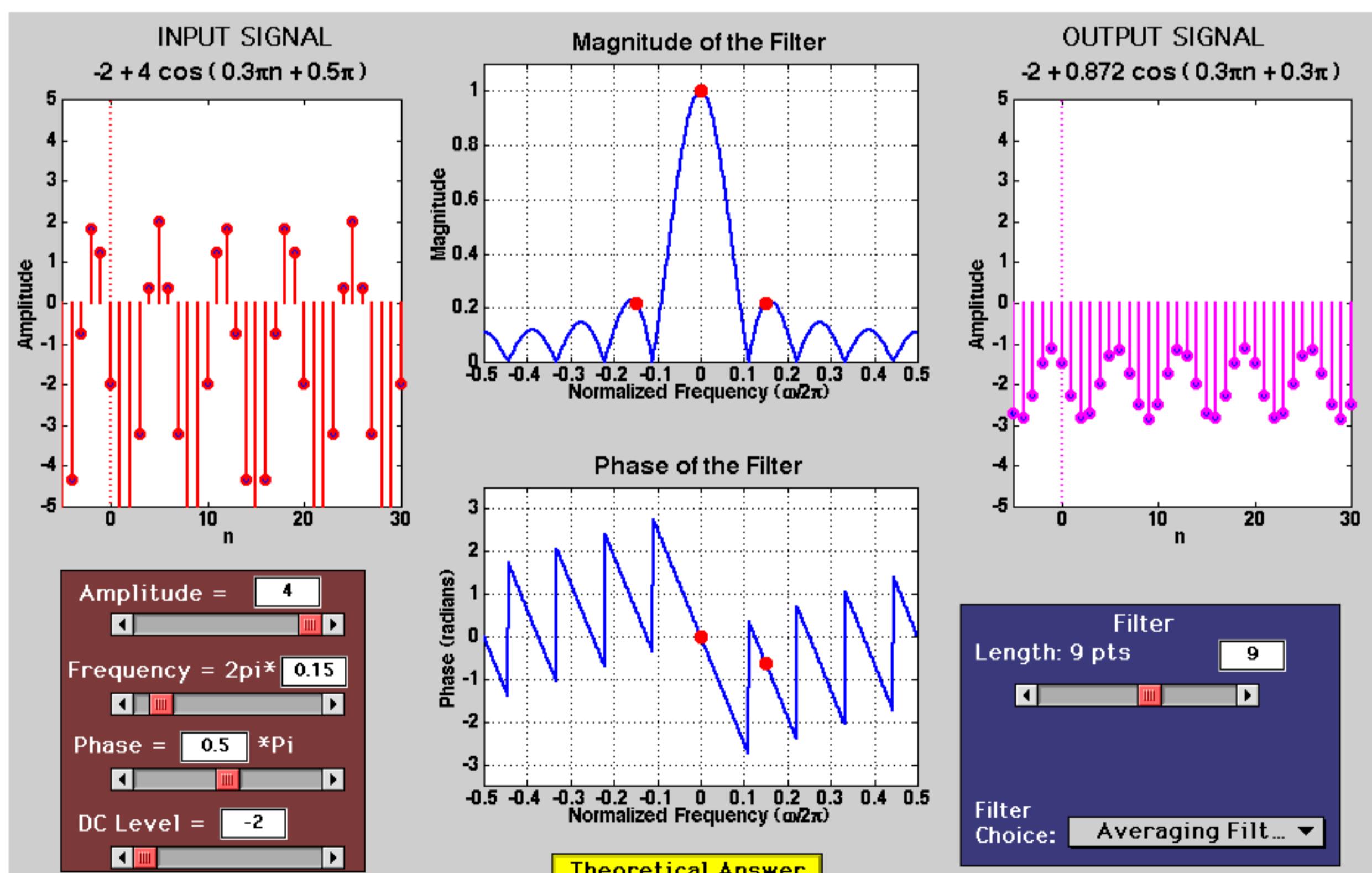
6.1 a) From left to right, the spectral lines correspond to the following time signals:

$$2e^{-j\frac{\pi}{2}} e^{-j0.3\pi n}, \quad 3e^{j\pi}, \text{ and}$$

$$2e^{j\frac{\pi}{2}} e^{j0.3\pi n}$$

$X[n]$  is the sum of these, or

$$X[n] = 4 \cos(0.3\pi n + \frac{\pi}{2}) - 3$$



Theoretical Answer

6.1

(b) The screen shot is included. Since the present version of the GUI only permits a DC value of -2, the input was set equal to

$$x[n] = -2 + 4 \cos(0.3\pi n + 0.5\pi)$$

The calculated output is

$$y[n] = -2 + 0.872 \cos(0.3\pi n + 0.3\pi)$$

$\nwarrow$  This would be -3 if the DC component of  $x[n]$  were equal to -3.

(c) Use the frequency response. Evaluate at  $\hat{\omega} = 0$  and at  $\hat{\omega} = 0.3\pi$ .

$$\mathcal{H}(\hat{\omega}) = \frac{1}{9} \sum_{k=0}^8 e^{-j\hat{\omega}k} = e^{-j4\hat{\omega}} \frac{\sin 9\hat{\omega}/2}{9 \sin \hat{\omega}/2}$$

$$\text{At } \hat{\omega} = 0 \quad \mathcal{H}(\hat{\omega}) = \frac{1}{9} \sum_{k=0}^8 e^{-j0} = \frac{1}{9}(9) = 1$$

At  $\hat{\omega} = 0.3\pi$

$$\mathcal{H}(\hat{\omega}) = e^{-j4(0.3\pi)} \frac{\sin(\frac{9}{2}(0.3\pi))}{9 \sin(\frac{1}{2}(0.3\pi))}$$

This term is negative

$$= e^{-j1.2\pi} \frac{\sin(1.35\pi)}{9 \sin(0.15\pi)} = 0.218 e^{-j0.2\pi}$$

The output is:

$$y[n] = \underbrace{(-3)\mathcal{H}(0)}_{-3} + \underbrace{4|\mathcal{H}(0.3\pi)|}_{0.872} \cos(0.3\pi n + 0.5\pi + \underbrace{\angle \mathcal{H}(0.3\pi)}_{0.3\pi})$$

## Prob 7.2

(a) All three are TRUE. The system is Linear, Time-Inv & Causal  
Linearity?

If  $x_1[n] \rightarrow y_1[n]$ , then  $y_1[n] = x_1[n] + 3x_1[n] + x_1[n]$

If  $x_2[n] \rightarrow y_2[n]$ , then  $y_2[n] = x_2[n] + 3x_2[n] + x_2[n]$

Define  $v[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$

Now, find the output when  $v[n]$  is the input. Call this output  $w[n]$ .

$$w[n] = v[n] + 3v[n-1] + v[n-2]$$

$$w[n] = (\alpha_1 x_1[n] + \alpha_2 x_2[n]) + 3(\alpha_1 x_1[n-1] + \alpha_2 x_2[n-1]) + (\alpha_1 x_1[n-2] + \alpha_2 x_2[n-2])$$

$$w[n] = \underbrace{\alpha_1 x_1[n] + 3\alpha_1 x_1[n-1] + \alpha_1 x_1[n-2]}_{\alpha_1 y_1[n]} + \underbrace{\alpha_2 x_2[n] + 3\alpha_2 x_2[n-1] + \alpha_2 x_2[n-2]}_{\alpha_2 y_2[n]}$$

Since  $w[n] = \alpha_1 y_1[n] + \alpha_2 y_2[n]$ , the system is LINEAR

Time-Invariant?

Assume  $x[n] \rightarrow y[n]$ . Define  $v[n] = x[n-n_0]$

Find the output when  $v[n]$  is the input. Call this output  $r[n]$ .

$$\begin{aligned} r[n] &= v[n] + 3v[n-1] + v[n-2] \\ &= x[n-n_0] + 3x[n-1-n_0] + x[n-2-n_0] \\ &= x[(n-n_0)] + 3x[(n-n_0)-1] + x[(n-n_0)-2] = y[n-n_0] \end{aligned}$$

Since  $r[n] = y[n-n_0]$ , the system is Time-Invariant.

Causal?

The system is causal because  $y[n] = x[n] + 3x[n-1] + x[n-2]$  uses  $x[n]$ ,  $x[n-1]$ , and  $x[n-2]$  to form  $y[n]$ .

(b) The filter coefficients are  $\{b_k\} = \{1, 3, 1\}$   $\sum_{k=0}^M b_k e^{-jk\hat{\omega}}$

$$H(\hat{\omega}) = 1 + 3e^{-j\hat{\omega}} + e^{-j\hat{\omega}}$$

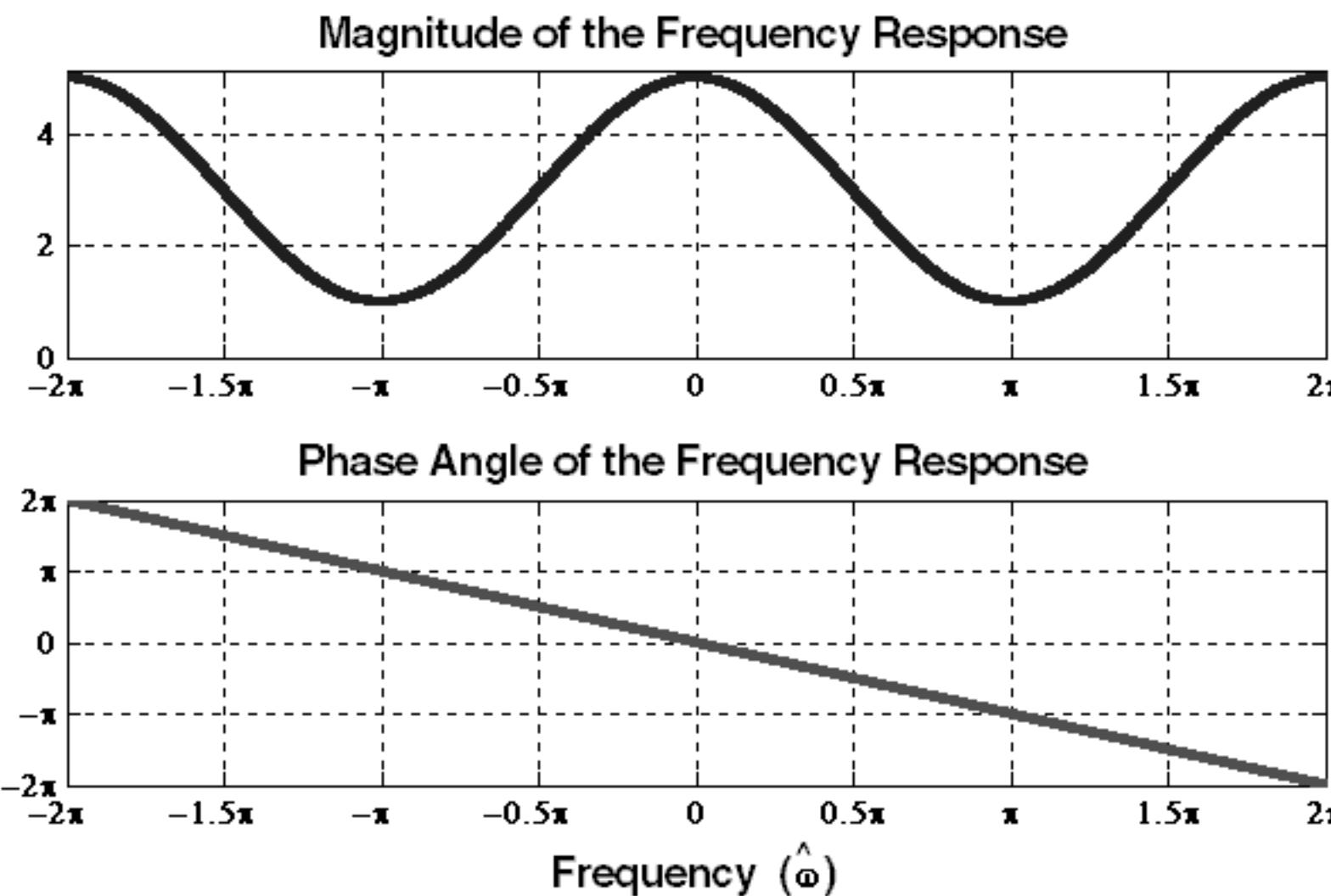
(c) Simplify first

$$H(\hat{\omega}) = e^{-j\hat{\omega}}(e^{j\hat{\omega}} + 3 + e^{-j\hat{\omega}}) = (3 + 2\cos\hat{\omega})e^{-j\hat{\omega}}$$

$$|H(\hat{\omega})| = 3 + 2\cos\hat{\omega} \quad \text{Magnitude}$$

$$\angle H(\hat{\omega}) = -\hat{\omega} \quad \text{phase}$$

## Prob 7.2 (cont)



$$(d) x_1[n] = 2\cos(0.75\pi n) = e^{j0.75\pi n} + e^{-j0.75\pi n}$$

Find the output when the input is  $e^{j0.75\pi n}$

$$\begin{aligned} y_{11}[n] &= \mathcal{H}(0.75\pi) e^{j0.75\pi n} \\ &= (3 + 2\cos(0.75\pi)) e^{-j0.75\pi} e^{j0.75\pi n} \\ &= (3 - \sqrt{2}) e^{j0.75\pi(n-1)} \end{aligned}$$

Find the output when the input is  $e^{-j0.75\pi n}$

$$\begin{aligned} y_{12}[n] &= \mathcal{H}(-0.75\pi) e^{-j0.75\pi n} \\ &= (3 + 2\cos(-0.75\pi)) e^{+j0.75\pi} e^{-j0.75\pi n} \\ &= (3 - \sqrt{2}) e^{-j0.75\pi(n-1)} \end{aligned}$$

By linearity, we can then add  $y_{11}[n] + y_{12}[n]$  to get  $y_1[n]$ .

$$\begin{aligned} y_1[n] &= (3 - \sqrt{2}) e^{j0.75\pi(n-1)} + (3 - \sqrt{2}) e^{-j0.75\pi(n-1)} \\ &= \underbrace{2(3 - \sqrt{2})}_{\approx 3.1716} \cos(0.75\pi(n-1)) \end{aligned}$$

$$(e) x_2[n] = 4 + 4\cos(0.75\pi(n-1))$$

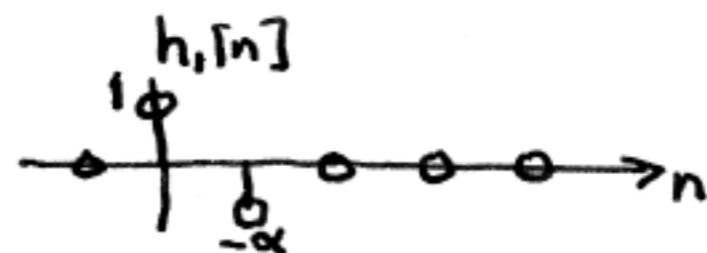
when the input is 4, the output is  $4 \cdot \mathcal{H}(0) = 4 \cdot 5 = 20$

when the input is  $4\cos(0.75\pi(n-1))$ , we use part (d) to get  $4(3 - \sqrt{2}) \cos(0.75\pi(n-2))$

$$\Rightarrow y_2[n] = 20 + 4(3 - \sqrt{2}) \cos(0.75\pi(n-2))$$

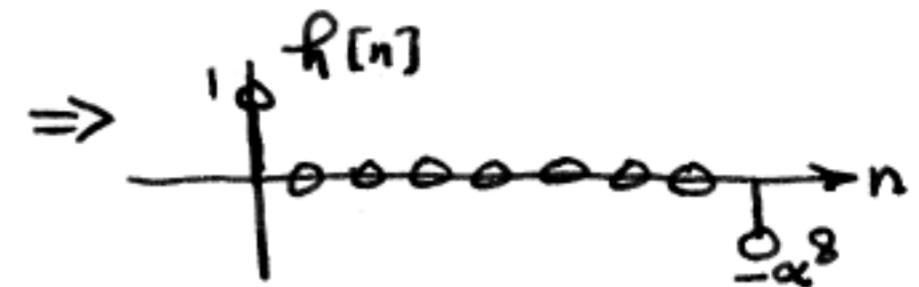
### Prob 7.3

(a)  $h[n] = \delta[n] - \alpha\delta[n-1]$



(b) Convolution:

$$\begin{array}{r} 1 \quad \alpha \quad \alpha^2 \quad \alpha^3 \quad \alpha^4 \quad \alpha^5 \quad \alpha^6 \quad \alpha^7 \\ \underline{1 \quad -\alpha} \\ 1 \quad \alpha \quad \alpha^2 \quad \alpha^3 \quad \alpha^4 \quad \alpha^5 \quad \alpha^6 \quad \alpha^7 \\ -\alpha \quad -\alpha^2 \quad -\alpha^3 \quad -\alpha^4 \quad -\alpha^5 \quad -\alpha^6 \quad -\alpha^7 \quad -\alpha^8 \\ \hline 1 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad -\alpha^8 \\ \uparrow \quad \quad \quad \quad \quad \uparrow \quad \quad \quad \quad \uparrow \\ n=0 \quad \quad \quad \quad \quad n=4 \quad \quad \quad \quad n=8 \end{array}$$



(c) For general values of L,  $h[n] = \delta[n] - \alpha^L\delta[n-L]$

(d) It is easy to convert  $h[n]$  back into a difference equation.

$$b_0 = 1, \quad b_L = -\alpha^L \Rightarrow y[n] = x[n] - \alpha^L x[n-L]$$

(e) If  $0 < \alpha < 1$ , then  $\alpha^L \rightarrow 0$  as  $L \rightarrow \infty$

so we should choose  $L = \infty$  to get  $y[n] = x[n]$ .

### Prob 7.4

(a)  $\mathcal{H}_1(\hat{\omega}) = 1 - e^{-j\hat{\omega}}$  because  $h_1[n] = \delta[n] - \delta[n-1] \Rightarrow b_0 = 1 \neq b_1 = -1$ .

(b) For  $h_2[n] = u[n] - u[n-8]$ ,  $b_k = 1$  for  $n=0, 1, 2, \dots, 7$

$$\mathcal{H}_2(\hat{\omega}) = \sum_{k=0}^7 b_k e^{jk\hat{\omega}} = 1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} + e^{-j4\hat{\omega}} + e^{-j5\hat{\omega}} + e^{-j6\hat{\omega}} + e^{-j7\hat{\omega}}$$

(c) Convolution:

$$\begin{array}{cccccccc} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ \hline 1 & -1 \\ \hline 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ \hline 1 & 0 & 0 & 0 & 0 & 0 & 0 & -1 \\ \uparrow & & \uparrow & & & & \uparrow \\ n=0 & & n=4 & & & & n=8 \end{array} \Rightarrow h[n] = \delta[n] - \delta[n-8]$$

(d)  $h[n] = \delta[n] - \delta[n-8] \Rightarrow b_0 = 1 \neq b_8 = -1$ .

$$\mathcal{H}(\hat{\omega}) = 1 - e^{-j8\hat{\omega}}$$

(e) Now do the multiplication of  $\mathcal{H}_1(\hat{\omega}) \mathcal{H}_2(\hat{\omega})$

$$(1 - e^{-j\hat{\omega}}) \mathcal{H}_2(\hat{\omega}) = 1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} + e^{-j4\hat{\omega}} + e^{-j5\hat{\omega}} + e^{-j6\hat{\omega}} + e^{-j7\hat{\omega}} + e^{-j8\hat{\omega}} + e^{-j9\hat{\omega}} - e^{-j10\hat{\omega}} - e^{-j11\hat{\omega}} - e^{-j12\hat{\omega}} - e^{-j13\hat{\omega}} - e^{-j14\hat{\omega}} - e^{-j15\hat{\omega}} - e^{-j16\hat{\omega}} - e^{-j17\hat{\omega}} - e^{-j18\hat{\omega}}$$

Therefore,  $(1 - e^{-j\hat{\omega}}) \mathcal{H}_2(\hat{\omega}) = 1 - e^{-j8\hat{\omega}}$  which matches part (d).

### Prob 7.5

(a)  $\mathcal{H}_1(\hat{\omega}) = e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} \Rightarrow b_1 = 1, b_2 = 1 \neq b_0 = 0$

$$y_1[n] = x_1[n-1] + x_2[n-2]$$

(b) For  $S_2$ ,  $b_0 = 1, b_1 = 0, b_2 = -1$ .  $\mathcal{H}_2(\hat{\omega}) = 1 - e^{-j2\hat{\omega}}$

For  $S_3$ ,  $b_4 = 5 \neq b_5 = 5 \Rightarrow \mathcal{H}_3(\hat{\omega}) = 5e^{-j4\hat{\omega}} + 5e^{-j5\hat{\omega}}$

(c) Overall Frequency Response is the product

$$\mathcal{H}_1(\hat{\omega}) \mathcal{H}_2(\hat{\omega}) \mathcal{H}_3(\hat{\omega}) = (e^{-j\hat{\omega}} + e^{-j2\hat{\omega}})(1 - e^{-j2\hat{\omega}})(5e^{-j4\hat{\omega}} + 5e^{-j5\hat{\omega}})$$

Simplify:  $= e^{-j\hat{\omega}} \cdot 5e^{-j4\hat{\omega}} (1 + e^{-j\hat{\omega}})(1 - e^{-j2\hat{\omega}})(1 + e^{-j\hat{\omega}})$   
 $= 5e^{-j5\hat{\omega}} (1 + 2e^{-j\hat{\omega}} - 2e^{-j3\hat{\omega}} - e^{-j4\hat{\omega}})$

(d) One more step with  $\mathcal{F}(w)$  and then we can get the  $\{b_k\}$

$$\mathcal{F}(\hat{\omega}) = \underbrace{5e^{-j5\hat{\omega}}}_{b_5} + \underbrace{10e^{-j6\hat{\omega}}}_{b_6} - \underbrace{10e^{-j8\hat{\omega}}}_{b_8} - \underbrace{5e^{-j9\hat{\omega}}}_{b_9}$$

$$y[n] = 5 \times [n-5] + 10 \times [n-6] - 10 \times [n-8] - 5 \times [n-9]$$

### Prob 7.6

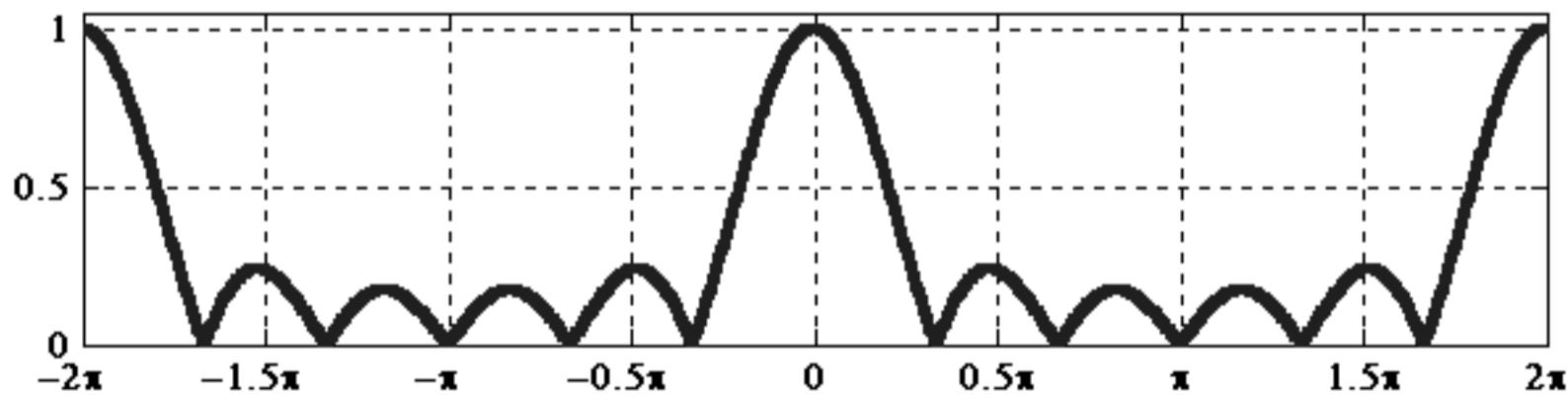
(a) From  $bb = \text{ones}(1, 6)/6$ ; we have  $b_k = 1/6$  for  $k=0, 1, 2, 3, 4, 5$

$$\begin{aligned} H(\hat{\omega}) &= \sum_{k=0}^5 \frac{1}{6} e^{-jk\hat{\omega}} = \frac{1}{6} \frac{1 - e^{-j6\hat{\omega}}}{1 - e^{-j\hat{\omega}}} \\ &= \frac{1}{6} \frac{e^{-j3\hat{\omega}}}{e^{-j\hat{\omega}/2}} \cdot \frac{e^{j3\hat{\omega}} - e^{-j3\hat{\omega}}}{e^{j\hat{\omega}/2} - e^{-j\hat{\omega}/2}} \\ &= e^{-j2.5\hat{\omega}} \frac{\sin(3\hat{\omega})}{6\sin(\hat{\omega}/2)} \end{aligned}$$

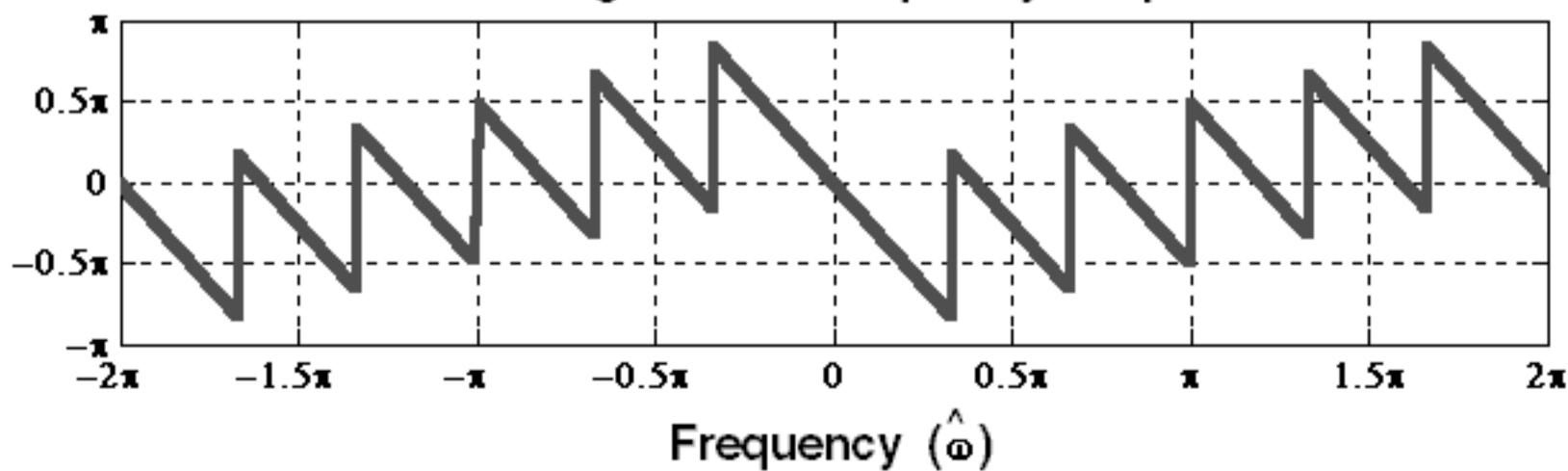
This part can  
be used to  
plot  $|H(\hat{\omega})|$

(b)

Magnitude of the Frequency Response



Phase Angle of the Frequency Response



## Prob 7.6

(c) The input signal values in  $x_n$  are  $[0, -3, 0, 3, 0, -3, 0, 3, 0, -3, \dots]$ .

Convolution:

$$\begin{array}{cccccccccccc}
 0 & -3 & 0 & 3 & 0 & -3 & 0 & 3 & 0 & -3 & 0 & 3 & 0 & -3 \\
 \frac{1}{6} & \frac{1}{6} & \frac{1}{6} & \frac{1}{6} & \frac{1}{6} & \frac{1}{6} & & & & & & & & \\
 \hline
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & -\frac{1}{2} & 0 & \frac{1}{2} & 0 & \\
 \hline
 0 & -\frac{1}{2} & -\frac{1}{2} & 0 & 0 & -\frac{1}{2} & -\frac{1}{2} & \frac{1}{2} & \frac{1}{2} & -\frac{1}{2} & -\frac{1}{2} & \frac{1}{2} & \frac{1}{2} & -\frac{1}{2} \dots \\
 \uparrow_{n=0} & & \uparrow_{n=5} & & & & & & \uparrow_{n=10} & & & & & &
 \end{array}$$

Notice that for  $n \geq 5$ , the period is 4, so the frequency is  $\frac{2\pi}{4} = \frac{\pi}{2}$ .

If we use the input  $3 \cos(\frac{\pi}{2}n + \frac{\pi}{2})$ , then the output for  $n \geq 5$  can be obtained by getting the magnitude & phase change from  $H(\frac{\pi}{2}) = e^{-j2.5\pi/2} \frac{\sin(3\pi/2)}{6\sin(\pi/4)} = e^{-j1.25\pi} \frac{-1}{3\sqrt{2}} = \frac{1}{3\sqrt{2}} e^{-j0.25\pi}$

$$\begin{aligned}
 \Rightarrow y[n] &= 3\left(\frac{1}{3\sqrt{2}}\right) \cos\left(\frac{\pi}{2}n + 0.5\pi - 0.25\pi\right) \quad \text{for } n \geq 5 \\
 &= \frac{1}{\sqrt{2}} \cos\left(\frac{\pi}{2}n + 0.25\pi\right) \quad \text{for } n \geq 5
 \end{aligned}$$

(d) The frequency response  $H(\hat{\omega}) = e^{-j2.5\hat{\omega}} \frac{\sin 3\hat{\omega}}{\sin(\hat{\omega}/2)}$  is zero

for  $\hat{\omega} = \frac{2\pi k}{6}$ ,  $k=1,2,3,4,5$ . Thus, all of these freqs are nulled.

$\hat{\omega} = \frac{\pi}{3}, \frac{2\pi}{3}, \pi, \frac{4\pi}{3} \text{ or } \frac{5\pi}{3}$  ← If the input frequency is one of these, the output will be zero, for  $n \geq 5$

### Prob 7.7

(a) To write the difference equation we need the  $b_k$ 's.

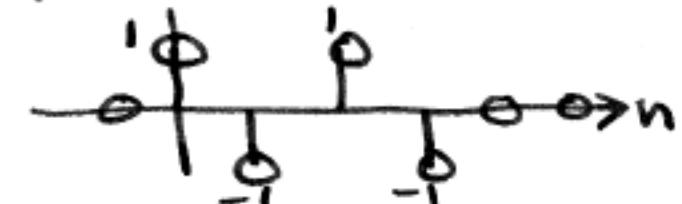
Expand  $\mathcal{H}(\hat{\omega})$  into a polynomial

$$\begin{aligned}\mathcal{H}(\hat{\omega}) &= (1 - e^{-j\hat{\omega}})(\underbrace{(1 - je^{-j\hat{\omega}})}_{(1 - je^{-j\hat{\omega}} + je^{-j\hat{\omega}} + e^{-j2\hat{\omega}})}(1 + je^{-j\hat{\omega}}) \\ &= 1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}} \Rightarrow b_0 = 1, b_1 = -1, b_2 = 1, b_3 = -1\end{aligned}$$

$$y[n] = x[n] - x[n-1] + x[n-2] - x[n-3]$$

(b) Use  $x[n] = \delta[n]$  to get the impulse response.

$$h[n] = \delta[n] - \delta[n-1] + \delta[n-2] - \delta[n-3]$$



(c)  $x[n] = Ae^{j\varphi} e^{j\hat{\omega}n}$  gives the output  $y[n] = \mathcal{H}(\hat{\omega}) Ae^{j\varphi} e^{j\hat{\omega}n}$   
so we need to find where  $\mathcal{H}(\hat{\omega}) = 0$ .

Use the FACTORED FORM. If one factor is zero then  $\mathcal{H}(\hat{\omega}) = 0$ .

$$1 - e^{-j\hat{\omega}} = 0 \Rightarrow \hat{\omega} = 0$$

$$1 - je^{-j\hat{\omega}} = 0 \Rightarrow e^{j\hat{\omega}} = j \Rightarrow \hat{\omega} = \pi/2$$

$$1 + je^{-j\hat{\omega}} = 0 \Rightarrow e^{j\hat{\omega}} = -j \Rightarrow \hat{\omega} = -\pi/2$$

$$\left. \begin{array}{l} \hat{\omega} = 0, \frac{\pi}{2}, -\frac{\pi}{2} \end{array} \right\}$$

(d) Work each term separately

$$x_1[n] = 1 \text{ is D.C. } \mathcal{H}(\hat{\omega})|_{\hat{\omega}=0} = 0 \Rightarrow y_1[n] = 0$$

$$x_2[n] = 2\delta[n-3]$$

use linearity & time-invariance to say that the output is  $h[n]$  shifted by 3 and doubled.

$$y_2[n] = 2h[n-3]$$

$$= 2\delta[n-3] - 2\delta[n-4] + 2\delta[n-5] - 2\delta[n-6]$$

$$x_3[n] = 7\cos(0.5\pi n). \text{ Since } \mathcal{H}(0.5\pi) = 0 \Rightarrow y_3[n] = 0$$

$$\therefore y[n] = y_1[n] + y_2[n] + y_3[n]$$

$$= 2h[n-3] \Rightarrow \text{plot}$$

