

RESERVE DESK

EE 3230 Solns PS # 4

SEP 27 1995 7/29/94

4.1) 4.14

i) $X(j\omega)$ real $\Rightarrow X(j\omega) = X^*(j\omega)$

since $X(j\omega) \rightarrow x(t)$ and $X^*(j\omega) \rightarrow x^*(-t)$

$$x(t) = x^*(-t) \text{ i.e. conjugate symm.}$$

since $x(t)$ real $\Rightarrow x(t)$ is even

c, f, h

ii) $X(j\omega)$ imag $\Rightarrow X(j\omega) = -X^*(j\omega)$

since $X(j\omega) \rightarrow x(t)$ and $-X^*(j\omega) \rightarrow -x^*(-t)$

$$x(t) = -x^*(-t) \text{ i.e. conjugate anti-symm.}$$

since $x(t)$ real $\Rightarrow x(t)$ is odd

b, e, g

iii) $X(0) = 0 \Rightarrow \int_{-\infty}^{\infty} x(t) e^{-j(0)t} dt = 0 \Rightarrow \int_{-\infty}^{\infty} x(t) dt = 0$

a, b, d, e, f, g, h

iv) $\int_{-\infty}^{\infty} X(j\omega) d\omega = 0 \Rightarrow \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{+j\omega(0)} d\omega = x(t=0) = 0$

b, c, d, e, g

v) $X(j\omega)$ periodic implies $x(t)$ consists of impulses only. This is because we know that a periodic signal in time can be expressed as a sum of impulses whose values are the Fourier series coefficients. So the spectrum would be impulses. Apply duality to get our case.

c, d

$$\text{vi) } e^{j\omega z} X(j\omega) = [e^{j\omega z} X(j\omega)]^* = e^{-j\omega z} X^*(j\omega)$$

$$e^{j\omega z} X(j\omega) \rightarrow x(t+z) \quad \text{and} \quad e^{-j\omega z} X^*(j\omega) \rightarrow x^*(-t-z)$$

$$\text{so } x(t+z) = x^*(-t-z) \quad \text{for some } z \neq 0$$

since $x(t)$ is real
 $x(t+z)$ is odd

a, d, f, g, h

$$\text{vii) } \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega = E \Rightarrow \int_{-\infty}^{\infty} |x(t)|^2 dt = E$$

by Parseval

a, b, c

$$\text{viii) } \int_{-\infty}^{\infty} \omega X(j\omega) d\omega = 0 \Rightarrow \frac{1}{2\pi} \int_{-\infty}^{\infty} j\omega X(j\omega) e^{j\omega t} d\omega \Rightarrow \left. \frac{dx(t)}{dt} \right|_{t=0} = 0$$

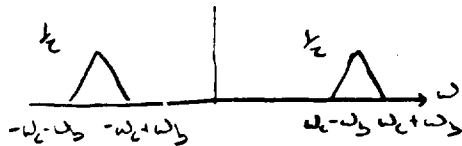
b, c, d, e, f

4.2) 4.18

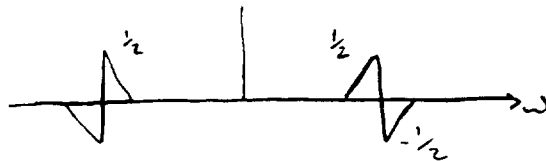
assume $x(j\omega) =$



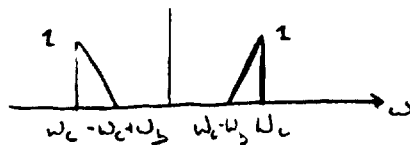
$V(j\omega) =$



$W(j\omega) =$

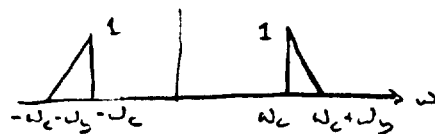


$V(j\omega) + W(j\omega) =$



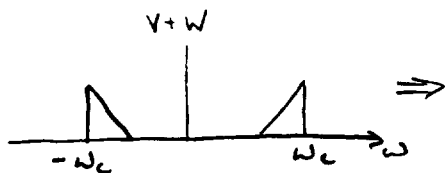
$= X_-(j\omega)e^{j\omega_c} + X_+(j\omega)e^{j\omega_c}$

$V(j\omega) - W(j\omega) =$



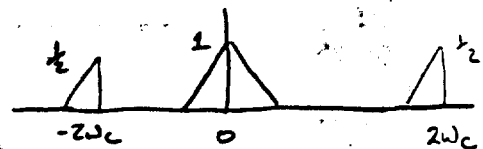
$= X_+(j\omega)e^{j\omega_c} + X_-(j\omega)e^{-j\omega_c}$

b)



\Rightarrow

demodulation shifts each sideband by $+\omega_c$ and $-\omega_c$ yielding:

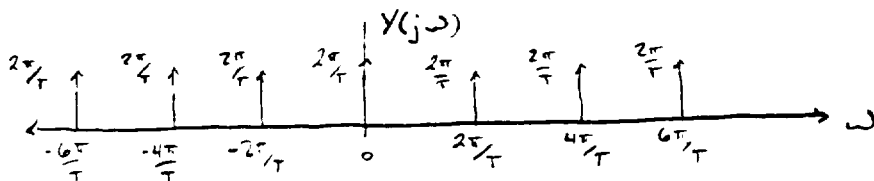


by low pass filtering, the component of interest (the one centered around $\omega=0$) can be removed

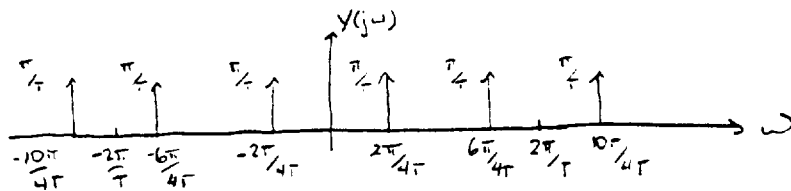
4.3) 4.23

$$Y(j\omega) = \frac{1}{2\pi} X(j\omega) * P(j\omega)$$

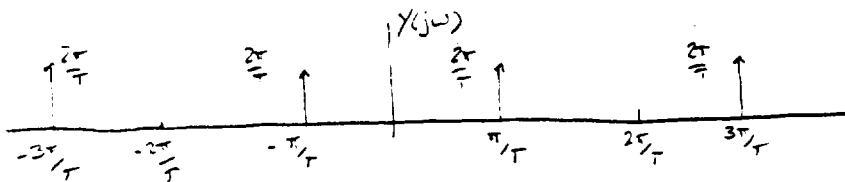
a) i) $x(t) = 1 \rightarrow X(j\omega) = 2\pi \delta(\omega)$



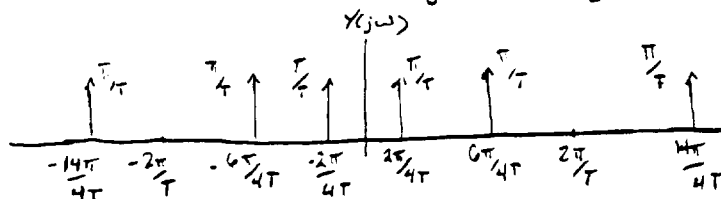
ii) $x(t) = \cos(\omega_s t / 4) \rightarrow X(j\omega) = 2\pi \left[\frac{1}{2} \delta(\omega - \frac{\pi}{2T}) + \frac{1}{2} \delta(\omega + \frac{\pi}{2T}) \right]$



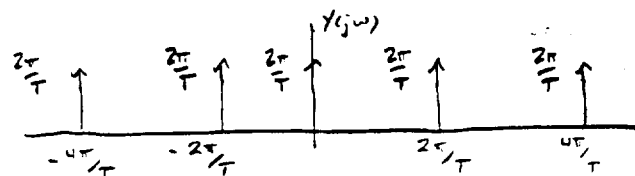
iii) $x(t) = \cos(\omega_s t / 2) \rightarrow X(j\omega) = \left[\frac{1}{2} \delta(\omega - \frac{\pi}{T}) + \frac{1}{2} \delta(\omega + \frac{\pi}{T}) \right] 2\pi$



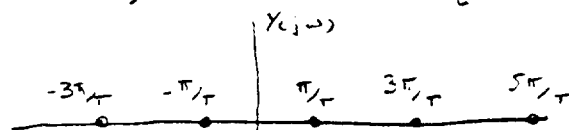
iv) $x(t) = \cos(3\omega_s t / 4) \rightarrow X(j\omega) = 2\pi \left[\frac{1}{2} \delta(\omega - \frac{6\pi}{4T}) + \frac{1}{2} \delta(\omega + \frac{6\pi}{4T}) \right]$



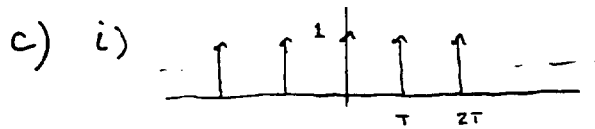
v) $x(t) = \cos(\omega_s t) \rightarrow X(j\omega) = \left[\frac{1}{2} \delta(\omega - \frac{2\pi}{T}) + \frac{1}{2} \delta(\omega + \frac{2\pi}{T}) \right] 2\pi$



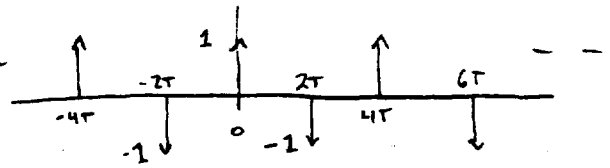
vi) $x(t) = \sin(\omega_s t / 2) \rightarrow X(j\omega) = \left[\frac{1}{2} \delta(\omega - \frac{\pi}{T}) - \frac{1}{2} \delta(\omega + \frac{\pi}{T}) \right] 2\pi$



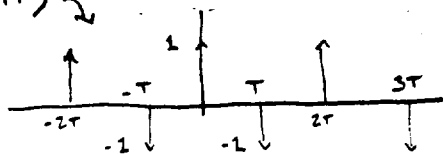
b) cases iii) - vi)



ii) $y(t) = \sum_{n=-\infty}^{\infty} \cos(\frac{\pi}{2T} nT) \delta(t-nT)$



iii) $\sum_{n=-\infty}^{\infty} \cos(\frac{\pi}{T} nT) \delta(t-nT)$



(see below)

4.4) 4.31

a) $H_{HP}(j\omega) = 1 - H_{LP}(j\omega) \Rightarrow h_{HP}(t) = \delta(t) - h_{LP}(t)$

b) $H_{BP}(j\omega) = H_{LP2}(j\omega) - H_{LP1}(j\omega) \Rightarrow h_{BP}(t) = h_{LP2}(t) - h_{LP1}(t)$

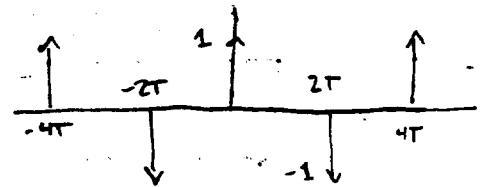
assume $\omega_{b2} > \omega_{b1}$

c) $H_{BS}(j\omega) = 1 - H_{BP}(j\omega) \Rightarrow h_{BS}(t) = \delta(t) - h_{BP}(t)$

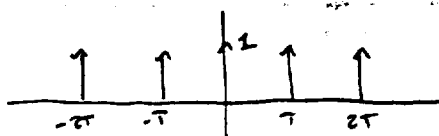
d) since each of the impulse responses integrate to a finite number they are all stable

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty \quad \forall \text{ the } h\text{'s in a-c}$$

4.23) c) cont w) $\sum_{n=-\infty}^{\infty} \cos(3\frac{\pi}{2T} nT) \delta(t-nT)$



v) $\sum_{n=-\infty}^{\infty} \cos(\frac{2\pi}{T} nT) \delta(t-nT)$



vi) $y(t) = 0$